# On Markov Channels 

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## 1．Introduction

The information theory originated by C．E．Shannon was extended to the case of finite memory channels by Hinchin ${ }^{3}$ ．Moreover the case of infinite memory channels was also studied by Adler ${ }^{1)}$ and others．However，among these generalized channels，there are not many which are useful for concrete models．In this paper，we define a Markov channel which is a natural genera－ lization of the Markov chain and useful to make examples of generalized chan－ nels．We show that it is determined by a pair of vector valued mappings，and give two representative examples of Markov channels and simulate how errors occur in these channels．We also show that a stationary stochastic state machine is a Markov channel and we can compute the capacity of this channel．

## 2．Notations

The input space and the output space are represented by the alphabet spaces $X=A^{I}$ and $Y=B^{I}$ respectively，where $A=\left\{a_{1}, a_{2}, \cdots \cdots, a_{k}\right\}$ and $B=\left\{b_{1}, b_{2}, \cdots \cdots, b_{l}\right\}$ are sets of finite letters（or states），and $I=\{0, \pm 1, \pm 2, \cdots\}$ is a set of all integers which represents the time，i．e．，$X$ and $Y$ are the infinite dimensional product spaces of the countable copies of finite measurable spaces $A$ and $B$ respectively． Let $\mathscr{X}$ and $\mathscr{Y}$ be Borel fields generated by all cylinder sets of $X$ and $Y$ res－ pectively．For the element $x$ in $X, x_{n}$ represents the $n^{\prime}$ th letter in $A$ ，and for $y$ in $Y, y_{n}$ represents the $n^{\prime}$ th letter in $B$ ．In the followings，we omit the explanation for $y$ as it is same as $x$ ．The notation $\left[x_{s}^{0}, x_{s+1}^{0}, \cdots, x_{t}^{0}\right]$ represents a thin cylinder on the time interval $(s, s+1, \cdots, t)$ ，i．e．，

$$
\left[x_{s}^{0}, x_{s+1}^{0}, \cdots, x_{t}^{0}\right]=\left\{x \in X: x_{n}=x_{n}^{0}, \quad n=s, s+1, \cdots, t\right\},
$$

where $s, t \in I(s<t)$ ．A $\sigma$－subfield of $\mathcal{Q}$ generated by all thin cylinders of the form $\left[x_{s}, x_{s+1}, \cdots, x_{t}\right]$ on the time interval $(s, s+1, \cdots, t)$ is denoted by $\otimes_{i=s}^{t} \mathscr{X}_{i}$ ， and a $\sigma$－subfield generated by all thin cylinders of the form $\left[x_{s}\right]$ is denoted by $\mathcal{X}_{s}$ and is called the time－s partition，as it is also seen as a finite partition of $X$ ．Let $2_{X}$ be a trivial partition of $X$ ，i．e．， $2_{X}=\{X\}$ ．A shift transformation $S$

[^0](or $T$ ) on $X$ (or $Y$ ) represents time advance and is defined by the formula ( $S x)_{n}$ $=x_{n+1}\left((T y)_{n}=y_{n+1}\right)$. The shift transformation $S$ (or $\left.T\right)$ is a measurable transformation on the measurable space $X$ (or $Y$ ). The input source is an $S$-invarariant probability measure $p$ on the measurable space $X$, i.e., which satisfies $p\left(S^{-1} E\right)$ $=p(E)$ for all $E \in \mathscr{X}$. The output source is defined similarly. We denote a set of all input sources by $\Pi$.

A channel from an input space $X$ to an output space $Y$ is a two variable numerical function $\nu_{x}(F)$ of $x \in X$ and $F \in \mathscr{Y}$ and satisfies the followings:
I. If we fix $x \in X$, then $\nu_{x}(\cdot)$ is a probability measure on $(Y, \mathscr{Y})$,
II. if we fix $F \in \mathscr{Y}$, then $\nu .(F)$ is a measurable function on $X$, and III. for all $x \in X$ and $F \in \mathscr{Y}, \nu_{S x}(F)=\nu_{x}\left(S^{-1} F\right)$.

An input source $p \in \Pi$ and a channel $\nu$ determine an output source $q(\cdot)$;

$$
q(F)=\int_{X} \nu_{x}(F) p(d x) \quad(F \in \mathscr{Y})
$$

Let $(X \otimes Y, \mathscr{P} \otimes \mathscr{Y})$ be the direct product measurable space of $(X, \mathscr{X})$ and $(Y, \mathscr{Y})$, then we call it a compound space, which is also an alphabet space $(A \otimes B)^{I}$. A shift transformation $S \otimes T$ on the compound space is defined by $(S \otimes$ $T)(x, y)=(S x, T y)$. A compound source is an $S \otimes T$-invariant probability measure on the compound space. An input source $p$ and a channel $\nu$ also determine a compound source $r$;

$$
r(G)=\int_{X} \nu_{x}\left(G_{x}\right) p(d x) \quad(G \in \mathscr{X} \otimes \mathscr{Y})
$$

where $G_{x}$ is an $x$-section of the set $G$. The sources $q(\cdot)$ and $r(\cdot)$ determined by an input source $p$ and a channel $\nu$ are sometimes denoted by $q(\cdot ; p, \nu)$ and $r(\cdot ;$ $p, \nu)$, respectively. We sometimes abbreviate $\nu_{x}\left(\left[y_{s}, y_{s+1}, \cdots, y_{t}\right]\right)$ to $\nu_{x}\left(y_{s} y_{s+1} \cdots y_{t}\right)$.

If there exists an integer $m$ such that $f(\cdot)=\nu\left(y_{1} y_{2} \cdots y_{n}\right)$ is a $\otimes_{i=1}^{n} \mathscr{X}^{0}$ imeasurable function for all positive integer $n$ and for all letters $y_{1}, y_{2}, \cdots, y_{n}$, and if $f(\cdot)$ is constant for $n$ smaller than $1-m$, then the channel $\nu$ is called a finite memory channel or an $m$-memory channel.

For any finite partitions $\mathscr{A}_{1}$ and $\mathscr{A}_{2}$ of $X$, we write

$$
\mathscr{A}_{1} \vee \mathscr{A}_{2}=\left\{A_{1} \cap A_{2}: A_{1} \in \mathscr{A}_{1}, A_{2} \in \mathscr{A}_{2}\right\},
$$

which is a new partition of $X$. Then the entropy $h_{p}(S)$ of the transformation $S$ relative to an input source $p \in \Pi$ is defined by

$$
h_{p}(S)=\lim _{n} \frac{1}{n} H\left(\mathscr{X}_{0} \vee S^{-1} \mathscr{X}_{0} \vee \cdots \vee S^{-n+1} \mathscr{X}_{0}\right),
$$

where $\mathscr{C}_{0}$ is the time-0 partition, and $H(\mathscr{A})$ is the sum $-\sum_{A \in \mathscr{A}} p(A) \log p(A)$.

If $q$ and $r$ an output source and a compound source respectively, then we can also define the entropy $h_{q}(T)$ and $h_{r}(S \otimes T)$ of the transformation $T$ relative to an output source $q$ and of the transformation $S \otimes T$ relative to a compound source $r$. Then the transmission rate $R_{p}(\nu)$ of a channel $\nu$ relative to an input source $p$ is given by

$$
R_{p}(\nu)=h_{p}(S)+h_{q}(T)-h_{r}(S \otimes T)
$$

where both $q(\cdot)=q(\cdot ; p, \nu)$ and $r(\cdot)=r(\cdot ; p, \nu)$ are determined by $p$ and $\nu$. The (stationary) capacity $C_{s}(\nu)$ of a channel $\nu$ is defined by

$$
C_{s}(\nu)=\sup _{p \in I} R_{p}(\nu) .
$$

## 3. Markov Channels

We begin with the definiton of a Markov channel.
Definition. A channel $\nu$ from the input space $X$ to the output space $Y$ is called an $m$-fold Markov channel if there exists a positive number $m$ such that
IV.

$$
\nu_{x}\left(y_{n} \mid y_{0} \cdots y_{n-1}\right)=\nu_{x}\left(y_{n} \mid y_{n-m} \cdots y_{n-1}\right)
$$

for all $x \in X$, for all $y_{1}, \cdots, y_{n} \in B$ and for all positive integer $n(n \geq m)$, where

$$
\nu_{x}\left(y_{s} \cdots y_{s+l} \mid y_{t} \cdots y_{s-1}\right)=\left\{\begin{aligned}
& \nu_{x}\left(y_{t} \cdots y_{s+l}\right) / \nu_{x}\left(y_{t} \cdots y_{s-1}\right) \\
& \cdots \text { if } \nu_{x}\left(y_{t} \cdots y_{s-1}\right) \neq 0 \\
& 0 \cdots \text { if } \nu_{x}\left(y_{t} \cdots y_{s-1}\right)=0 .
\end{aligned}\right.
$$

The errors which occur in the $m$-fold Markov channel depend on the only $m$ steps in the past.

Next let us see that the Markov channel is characterized by a pair of measurable mappings. Now for every element $x$ in $X$, we consider a stochastic vector on $B^{m}$

$$
Q(x)=\left(q_{\bar{b}}(x)\right) \quad\left(\bar{b} \in B^{m}\right)
$$

and we also consider a stochastic matrix

$$
\widetilde{Q}(x)=\left\langle\tilde{q}_{b} \mid \bar{b}(x)\right\rangle,
$$

whose row is indexed by $b$ in $B$ and whose column is indexed by $\vec{b}$ in $B^{m}$. Moreover we assume that $q_{\bar{b}}(x)$ and $\tilde{q}_{b \mid \bar{b}}(x)$ are measurable functions of $x$ on ( $X, \mathscr{X}$ ), and assume that

$$
\begin{equation*}
q_{y_{n-m+1} \cdots y_{n}}(S x)=\sum_{y_{n-m} \in B} \tilde{q}_{y_{n} \mid y_{n-m} \cdots y_{n-1}}(x) q_{y_{n-m} \cdots y_{n-1}}(x) . \tag{1}
\end{equation*}
$$

The pair $(Q(x), \widetilde{Q}(x))$ is called a stochastic pair.

Theorem 1. Let $(Q(x), \widetilde{Q}(x))$ be a stochastic pair. Then it defines an $m$-fold Markov channel. Conversely every m-fold Markov channel defines a stochastic pair.

Proof. For a stochastic pair $(Q(x), \widetilde{Q}(x))$, we can construct an $m$-fold Markov channel by the following proceedure:
We put

$$
\nu_{x}\left(y_{1} \cdots y_{n}\right)= \begin{cases}q_{y_{1} \cdots y_{m}}(x) \Pi I_{i=1}^{n-m} \tilde{q}_{y_{m+i} \mid y_{i} \cdots y_{m+i-1}}\left(S^{i-1} x\right) \cdots \text { if } n-m>0 \\ \sum_{y_{n} \cdots y_{m}} q_{y_{1}} \cdots y_{m}(x) & \cdots \text { if } n-m \leq 0\end{cases}
$$

Assumming $n-m>0$, we see

$$
\begin{aligned}
\sum_{y_{n} \in B} \nu_{x}\left(y_{1} \cdots y_{n}\right) & =\sum_{y_{n} \in B} q_{y_{1} \cdots y_{m}}(x) \Pi_{i=1}^{n-m} \tilde{q}_{y_{m+i} \mid y_{i} \cdots y_{m+i-1}}(x) \\
& =q_{y_{1} \cdots y_{m}}(x) \Pi_{i=1}^{n-m-1} \tilde{q}_{y_{m+i} \mid y_{i} \cdots y_{m+i-1}}\left(S^{i-1} x\right) \\
& =\nu_{x}\left(y_{1} \cdots y_{n-1}\right) .
\end{aligned}
$$

As a similar equality also holds in case of $n-m \leq 0$, we conclude that $\nu_{x}(\cdot)$ can be extended to a probability measure on $\otimes_{i=1}^{\infty} \mathscr{V}_{i}$ by the Kolmogorov extension theorem. Next, for any thin cylinder $\left[y_{l}, \cdots, y_{n}\right]$ in $Y(n \geq l \geq 1, n>m)$,

$$
\begin{aligned}
\nu_{x}\left(y_{l} \cdots y_{n}\right) & =\sum_{y_{1} \cdots y_{l-1}} \nu_{x}\left(y_{1} \cdots y_{m}\right) \\
& =\left.\sum_{y_{1} \cdots y_{l-1}} q_{y_{1} \cdots y_{m}}(x) \prod_{i=1}^{n-m} \tilde{q}_{y_{m+i}}\right|_{y_{i} \cdots y_{m+i-1}}\left(S^{i-1} x\right) \\
& =\left.\sum_{y_{2} \cdots y_{l-1}} q_{y_{2} \cdots y_{m+1}}(S x) \Pi_{i=2}^{n-m} \tilde{q}_{y_{m+i} \mid}\right|_{y_{i} \cdots y_{m+i-1}}\left(S^{i-1} x\right) \\
& =\cdots \cdots \cdots \\
& = \begin{cases}q_{y_{l} \cdots y_{m+1}}(S l x) \prod_{i=l}^{n-m} \tilde{q}_{y_{m+i}} \mid y_{i \cdots y_{m+i-1}\left(S^{i-1} x\right) \cdots \text { if } n-m \geq l}^{\sum_{y_{n-m} \cdots y_{l-1}} q_{y_{n-m} \cdots y_{n-1}}\left(S_{n-m} x\right)} \quad \cdots \text { if } n-m<l .\end{cases}
\end{aligned}
$$

In both cases of $n>m$ and $n \leq m$, we get

$$
\begin{equation*}
\nu_{x}\left(y_{l} \cdots y_{n}\right)=\nu_{S} l_{x}\left(T^{l}\left[y_{l} \cdots y_{n}\right]\right) \tag{2}
\end{equation*}
$$

Replacing $l$ in (2) by $l-1$, we get

$$
\nu_{S x}\left(T\left[y_{l} \cdots y_{n}\right]\right)=\nu_{S l-1(S x)}\left(T^{l-1} T\left[y_{l}, \cdots, y_{n}\right]\right)
$$

$$
=\nu_{S} t_{x}\left(T^{l}\left[y_{l}, \cdots, y_{n}\right]\right)
$$

as $T\left[y_{l}, \cdots, y_{n}\right]$ is a thin cylinder on the time interval $(l-1, \cdots, n-1)$. Hence

$$
\nu_{x}\left(y_{l} \cdots y_{n}\right)=\nu_{S x}\left(T\left[y_{l} \cdots y_{n}\right]\right)
$$

holds. We extend $\nu_{x}(\cdot)$ from $\otimes_{i=1}^{\infty} \mathscr{V}_{i}$ to $\mathscr{Z}=\otimes_{i=-\infty}^{\infty} \mathscr{Y} i$, by the formula

$$
\nu_{x}\left(y_{-l} \cdots y_{n}\right)=\nu_{S^{-}} l_{x}\left(T^{-l}\left[y_{-l}, \cdots, y_{n}\right]\right)
$$

Then $\nu$ satisfies the condition III of the channel. For any $\left[y_{l}, \cdots, y_{n}\right], \nu .\left(y_{l} \cdots y_{n}\right)$ is $\mathscr{Q}^{\circ}$-measurable on $X$, as this is defined by the product of the measurable functions $q \bar{b}(x)$ and $\tilde{q}_{b_{1} \bar{b}}(x)$. And for any measurable set $F(\in \mathbb{Z}), \nu .(F)$ is also measurable, as $F$ can be approximated by a countable sequence of cylinder sets with the measure $\nu_{x}(\cdot)$. Therefore the condition II is also clear.

Conversely let $\nu$ be an $m$-fold Markov channel. We define a stochastic pair $(Q(x), \widetilde{Q}(x))$ by the following;

$$
q_{y_{1}}(x)=\nu_{x}\left(y_{1}\right)
$$

and

$$
\tilde{q}_{y_{m+1} \mid \bar{b}}(x)=\nu_{x}\left(y_{m+1} \mid \bar{b}\right)
$$

where $\bar{b}=\left[y_{1}, y_{2}, \cdots y_{m}\right]$. Clearly the pair is well defined by the condition III of the channel.

$$
\begin{aligned}
& \sum_{y_{1} \in B} \tilde{q}_{y_{m+1} \mid y_{1} \cdots y_{m}}(x) q_{y_{1} \cdots y_{m}}(x)=\sum_{y_{1} \in B} \nu_{x}\left(y_{1} \cdots y_{m}\right) \nu_{x}\left(y_{m+1} \mid y_{1} \cdots y_{m}\right) \\
& \quad=\sum_{y_{1} \in B} \nu_{x}\left(y_{1} \cdots y_{m+1}\right)=\nu_{x}\left(y_{2} \cdots y_{m_{+1}}\right)=\nu_{S x}\left(T\left(y_{2} \cdots y_{m+1}\right)\right) \\
& \quad=q_{y_{2} \cdots y_{m+1}}(S x)
\end{aligned}
$$

The above chain of formulae shows that the pair satisfies the condition of the stochastic pair.
Q.E.D.

This theorem tells us that any Markov channel is determined by a suit of mappings, which resembles to the fact that a memoryless channel (cf. 2) p. 154) is determined by a stochastic matrix. Let $\left(c_{a b}\right)$ be a stochastic matrix which determines a memoryless channel, then taking $q_{b}(x)=c_{x_{1} b}$ and $\tilde{q}_{b_{1} b}(x)=c_{x_{2} b}$, we see that the memoryless channel is a kind of Markov channels.

## 4. Examples

We put $A=B=\{0,1\}$, i. e., we consider a binary case. For a stochastic pair, let us choose

$$
\begin{aligned}
& Q(x)=\left\{\begin{array}{cl}
\binom{p}{1-p}=Q(0) & \cdots \text { if } x_{0}=0, \\
\binom{1-p}{p}=Q(1) & \cdots \text { if } x_{0}=1,
\end{array}\right.
\end{aligned}
$$

and assume that $\widetilde{Q}(00) Q(0)=Q(0) \quad$ i. e., $p=(r-1) /(r+q-2) . \quad$ Then we find that $\widetilde{Q}(11) Q(1)=Q(1), \widetilde{Q}(01) Q(0)=Q(1)$ and $\widetilde{Q}(10) Q(1)=Q(0)$. Hence $(Q(x), \widetilde{Q}(x))$ is a stochastic pair, so that we can define a Markov channel, which we call an additive noise channel. Errors in this channel occur as shown in Fig. 1. Fig. 2 is a result of a simple simulation in the case of $q=9 / 10, r=1 / 2$ and $p=5 / 7$.


Fig. 1


Fig. 2

Next we give another example;

$$
\begin{aligned}
& Q(x)=\left(\begin{array}{l}
\left.\left.\frac{1}{\frac{1}{2}}\right)^{2}\right)=Q \\
Q(x)=\left\{\begin{array}{cc}
\left(\begin{array}{cc}
p & 1-p \\
1-p & p
\end{array}\right)=Q(0) & \text { if } x_{1}=0, \\
\left(\begin{array}{cc}
q & 1-q \\
1-q & q
\end{array}\right)=Q(1) & \text { if } x_{1}=1 .
\end{array}\right.
\end{array} .\right.
\end{aligned}
$$

Then we can also define a Markov channel as shown in Fig. 3, which we call a stationary stochastic state machine channel (SSSM channel). Fig. 4 is a result of simulation when $p=3 / 4$ and $q=1 / 4$. In Fig. 4 , the output letters seem to have little correlation with the input letters, but we can show later that the capacity of this channel is not zero, therefore, the transmission of the information is possible through this noisy channel by the Shannon-Hinchin theorem.


$x_{1}=1$


Fig. 3


Fig. 4

## 5. Capacity of SSSM Channels

The second example of the previous section can be extended to a more general case, i. e., the letters need not be binary. Let $Q(x)$ be independent of $x \in X$. And we assume that $\widetilde{Q}(x)$ is only dependent on $x$, hence we can write $\widetilde{Q}(x)=\widetilde{Q}\left(x_{1}\right)$, and assume that $\widetilde{Q}\left(x_{1}\right) Q=Q$ for all $x \in A$, where $Q=Q(x)$. We call this generalized one an SSSM channel also. We can compute the capacity
$C_{s}(\nu)$ of an SSSM channel $\nu$. First, we give a proposition in a more general situation.

Proposition 1. Let $\nu$ be a finite memory Markov channel, then

$$
R_{p}=h_{q}(T)+\int_{X} \sum_{y_{1} \cdots y_{m+1}} q_{y_{1} \cdots y_{m}}(x) \eta\left(\tilde{q}_{y_{m+1} \mid y_{1} \cdots y_{m}}(x)\right) p(d x)
$$

where $\eta(t)=-t \log t$.
Proof. By the definition of $R_{p}$,

Here we assume that $\nu$ is $m$-fold Markovian and of $m$-memory with the common integer $m$, which does not lose generality.

$$
\begin{aligned}
R_{p} & =h_{q}(T)+h_{p}(S)+\left(\lim _{n} \frac{1}{n} \sum_{x_{1-m} \cdots x_{n}} \sum_{y_{1} \cdots y_{n}} \eta\left(p\left(x_{1-m} \cdots x_{n}\right) \nu_{x}\left(y_{1} \cdots y_{n}\right)\right)\right. \\
& =h_{q}(T)+h_{p}(S)+\lim _{n} \frac{1}{n} \sum_{x_{1} \cdots m x_{n}} \sum_{y_{1} \cdots y_{n}} \eta\left(p\left(x_{1-m} \cdots x_{n}\right) q_{y_{1} \cdots y_{m}}(x)\right.
\end{aligned}
$$

$$
\left.\cdot \prod_{i=1}^{m-n} \tilde{q}_{y_{i+m} \mid} \mid y_{i} \cdots y_{i+m-1}(x)\right)
$$

$$
=h_{q}(T)+h_{p}(S)
$$

$$
\left.+\lim _{n} \frac{1}{n_{x_{1-m}}} \sum_{n} \sum_{y_{1} \cdots y_{n}} p\left(x_{1-m} \cdots x_{n}\right) q_{y_{1} \cdots y_{m}}(x) \Pi_{i=1}^{n-m} \tilde{q}_{y_{i+m} \mid} \right\rvert\, y_{i} \cdots y_{i+m-1}(x) \times
$$

$$
\times\left(\log p\left(x_{1-m} \cdots x_{n}\right)+\log q_{y_{1} \cdots y_{m}}(x)+\sum_{j=1}^{n-m} \log \tilde{q}_{y_{j+m} \mid y_{j} \cdots y_{j+m-1}}(x)\right)
$$

$$
\begin{aligned}
& R_{p}=h_{p}(S)+h_{q}(T)-h_{r}(S \otimes T) \\
& =h_{q}(T)+h_{p}(S)-\lim _{n} \frac{1}{n} H\left(\mathbb{Z}_{0} \otimes \mathscr{Y} 0 \vee S^{-1} \mathscr{X}_{0} \otimes T^{-1} \mathscr{V}_{0} V\right. \\
& \left.\cdots \vee S^{-n+1} \mathscr{X}_{0} \otimes T^{-n+1} \mathscr{V}_{0}\right) \\
& =h_{q}(T)+h_{p}(S) \\
& -\lim _{n} \frac{1}{n} H\left(S^{m-1} \mathscr{X}_{0} \otimes 2_{Y} \vee S^{m} \mathscr{X}_{0} \otimes 2_{Y} V\right. \\
& \cdots \vee S \mathscr{P}_{0} \vee 2_{Y} \vee \mathscr{X}_{0} \otimes \mathscr{Y}_{0} \vee S^{-1} \mathscr{P}_{0} \otimes T^{-1} \mathscr{Y}_{0} \vee \\
& \left.\cdots \vee S^{-n+1} \mathscr{P}_{0} \otimes T^{-n+1} \mathscr{Y}_{0}\right) \\
& =h_{q}(T)+h_{p}(S)+\lim _{n} \frac{1}{n} \sum_{x_{1-m} \cdots x_{n}} \sum_{y_{1} \cdots y_{n}} n\left(\int \nu_{\substack{x_{1}-m \\
x_{1}-x_{n}}}\left(y_{1} \cdots y_{n}\right) p(d x)\right) .
\end{aligned}
$$

$$
\begin{aligned}
& =h_{q}(T)+\lim _{n} \frac{1}{n} \sum_{x_{1-m} \cdots x_{n} y_{1} \cdots y_{n}} p\left(x_{1-m} \cdots x_{n}\right) q_{y_{1}} \cdots y_{m}(x) \\
& \left.\cdot \Pi_{i=1}^{n-m} \tilde{q}_{y_{i+m} \mid y_{i} \cdots y_{i+m-1}}(x)\right)\left(\sum_{j=1}^{n-m} \log \tilde{q}_{\left.y_{j+m} \mid y_{j \cdots y_{j+m-1}}(x)\right)}\right. \\
& =h_{q}(T)+\lim _{n} \frac{1}{n}\left\{\sum_{j=1}^{n-m} \sum_{x_{1-m} \cdots x_{n}} \sum_{y_{1} \cdots y_{n}} p\left(x_{1-m} \cdots x_{n}\right) \nu_{x}\left(y_{1} \cdots y_{n}\right) \log \tilde{q}_{y_{j+m}: y_{j} \cdots y_{m+j-1}}(x)\right\} \\
& =h_{q}(T)+\lim _{n} \frac{1}{n}\left\{\sum _ { j = 1 } ^ { n - m } \sum _ { x _ { j - m } \cdots x _ { j + m } } \sum _ { y _ { j } \cdots y _ { j + m - 1 } } p \left(x_{\left.j_{-m} \cdots x_{j+m}\right)}\right.\right. \\
& \text { - } \left.\nu_{x}\left(y_{j} \cdots y_{j_{+m-1}}\right) \log \tilde{q}_{j_{j+m} \mid y_{1} \cdots y_{m+j-1}}(x)\right\} \\
& =h_{q}(T)+\lim _{\pi} \frac{1}{n}\left\{\sum _ { j = 1 } ^ { n - m } H \left(S^{j+m} 2_{X} \otimes T^{j+m} \mathscr{V}_{o}\right.\right. \\
& \left.\left.V_{i=j-m}^{j-1} S^{i} \mathscr{C}_{0} \otimes T^{i} 2_{Y} \vee V_{k=j}^{j+m-1} S^{k} \mathscr{C}_{0} \otimes T^{k} \not \mathscr{Z}_{0}\right)\right\} \\
& =h_{q}(T)+\int_{X} \sum_{y_{1} \cdots y_{m+1}} q_{y_{1} \cdots y_{m}}(x) \eta\left(\tilde{q}_{y_{m+1} \mid} y_{1} \cdots y_{m}(x)\right\rangle p(d x),
\end{aligned}
$$

which completes the proof, where $H(\mathscr{A} \mid \mathscr{S})$ means $H(\mathscr{A} \vee \mathscr{B})-H(\mathscr{S})$.
Q. E. D.

Next we give another proposition:
Proposition 2. Let $\nu$ be an SSSM channel defined by $Q=\left(q_{b}\right)$ and $\widetilde{Q}\left(x_{1}\right)=$ ( $\tilde{q}_{b \mid b^{\prime}}\left(x_{1}\right)$ ). If $p_{0}$ is a Bernoulli input source, i.e., time independent probability measure on $X=A^{I}$, then the output source $q_{0}(\cdot)=q\left(\cdot ; p_{0}, \nu\right)$ is a Markov measure on $Y=B^{I}$.

Proof. Let us prove that

$$
\begin{equation*}
q_{0}\left(\left[y_{n}\right] \mid\left[y_{1} \cdots y_{n-1}\right]\right)=q_{0}\left(\left[y_{n}\right] \mid\left[y_{n-1}\right]\right) . \tag{3}
\end{equation*}
$$

We abride the brackets [] in the following formulae. Then,

$$
\begin{aligned}
& q_{n}\left(y_{n} \mid y_{1} \cdots y_{n-1}\right)=q_{0}\left(y_{1} \cdots y_{n-1} y_{n}\right) / q_{0}\left(y_{1} \cdots y_{n-1}\right) \\
& \left.=q\left(y_{1}\right) \tilde{q}_{y_{21} \mid y_{1}\left(x_{2}\right) \cdots \tilde{q}_{y_{n} \mid y_{n-1}}\left(x_{n}\right) p\left(x_{1}\right) \cdots p\left(x_{n}\right) /\left(q\left(y_{1}\right) \tilde{q}_{y_{2} \mid y_{1}\left(x_{2}\right)}\right.} \quad \cdots \tilde{q}_{y_{n-1} \mid} \mid y_{n-2}\left(x_{n-1}\right) p\left(x_{1}\right) \cdots p\left(x_{n-1}\right)\right) \\
& =\tilde{q}_{y_{n} \mid y_{n-1}\left(x_{n}\right) p\left(x_{n}\right)=q_{0}\left(y_{n} \mid y_{n-1}\right),}
\end{aligned}
$$

which proves the equality (3).
Q. E. D.

Theorem 2. Let $\nu$ be an SSSM channel, then the capacity of $\nu$ is given by

$$
C_{s}(\nu)=\sup _{p_{0}}\left\{\sum_{a_{2} b_{1} b_{2}} q_{b_{1}} \tilde{q}_{b_{2} \mid b_{1}}\left(a_{2}\right) p_{0}\left(a_{2}\right) \log \left(\tilde{q}_{b_{2} \mid b_{1}}\left(a_{2}\right)\right) /\left(\sum_{a_{2^{\prime}}} \tilde{q}_{b_{2} \mid b_{1}}\left(a_{2}^{\prime}\right) p_{0}\left(a_{2}^{\prime}\right)\right)\right\},
$$

where $p_{0}$ moves on all probability measures on $A$.
Proof. Let $p$ be any input source, then by Proposition 1,

$$
R_{p}=h_{q}(T)+\sum_{y_{1}} q_{y_{1}} \eta\left(\tilde{q}_{y_{2} \mid y_{1}}\left(x_{2}\right)\right) p\left(x_{2}\right) .
$$

The following inequality and equality are well known (cf. 2) p. 82 and p.78) :

$$
h_{q}(T)=\lim _{n} H\left(S \mathscr{Y}_{0} \mid \vee_{i=0}^{n} S^{-i} \mathscr{Y}_{0}\right) \leqq H\left(S \mathscr{Y}_{0} \mid \mathscr{Y}_{0}\right) .
$$

Thus,

$$
h_{q}(T) \leq \sum_{y_{0}} q\left(y_{0}\right) \eta\left(q\left(y_{1} \mid y_{0}\right)\right)=\sum_{y_{0}} q_{y_{0}} \eta\left(\sum_{x_{1}} q_{y_{1} \mid y_{0}}\left(x_{1}\right) p\left(x_{1}\right)\right) .
$$

Now let $p_{0}$ be a Bernoulli source given by

$$
p_{0}\left(x_{1} \cdots x_{n}\right)=p\left(x_{1}\right) p\left(x_{2}\right) \cdots p\left(x_{n}\right),
$$

then by Proposition 2, denoting $q_{0}(\cdot)=q\left(\cdot ; p_{0}, \nu\right)$,

$$
H\left(S^{-1} \mathscr{Y}_{0} \mid \mathscr{Y}_{0}\right)=h_{q_{0}}(T)
$$

Therefore

$$
\begin{aligned}
R_{p} & =h_{q}(T)+\sum_{y_{1}} q_{y_{1}} \eta\left(\tilde{q}_{y_{2} \mid y_{1}}\left(x_{2}\right)\right) p\left(x_{2}\right) \\
& \leq h_{q_{0}}(T)+\sum_{y_{1}} q_{y_{1}} \eta\left(\tilde{q}_{y_{2} \mid y_{1}}\left(x_{2}\right)\right) p_{0}\left(x_{2}\right) \quad\left(=R_{p_{0}}\right) \\
& =\sum_{a_{2} b_{1} b_{2}} q_{b_{1}} \tilde{q}_{b_{2} \mid b_{1}}\left(a_{2}\right) p_{0}\left(a_{2}\right) \log \left(\tilde{q}_{b_{2} \mid b_{1}}\left(a_{2}\right)\right) /\left(\sum_{a_{2^{\prime}}} \tilde{q}_{b_{2} \mid b_{1}( }\left(a_{2^{\prime}}\right) p_{0}\left(a_{2}^{\prime}\right)\right),
\end{aligned}
$$

which proves the result.
Q. E. D.

Therefore we can easily compute the capacity of the channel shown in Fig. 2 of the previous section, which is about 0.1886 bit/step.

## Reference

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